Calculation Approach of Effective Permittivity at the Interface of Dielectric Media in FDTD

Kyeong-Sik Min and Manh-Dat Vu

Department of Radio Sciences and Engineering, College of Sciences and Engineering, Korea Maritime University, #1 Dongsam-Dong Yeongdo-Ku Busan, 606-791, Korea E-mail: ksmin@mail.hhu.ac.kr, Tel: +82-51-410-4425

Abstract

The solutions of effective permittivities at the dielectric interface in the Yee's FDTD algorithm are theoretically investigated by using the boundary conditions. At the interface of two dielectric media, the specific values of these effective permittivities must be determined in order to express the proper solutions from one region into the next with the high accuracy in the FDTD analysis.

In electric properties across the discontinuous distribution of two dielectric media, the effective permittivities are chosen to preserve the relations between the field components on adjacent sides of the boundaries. The results of numerical analysis in examples have verified the reasonable agreement between theoretical results and numerical experiments.

I. INTRODUCTION

With the rapid development of computer techniques and information technologies in recent decades, Finite- Difference Time-Domain (FDTD) becomes a popular and efficient computational electrodynamics modeling technique. The FDTD technique can allow the user to specify the material at all points within the computational domain [1]. In dealing with complex shape objects, where the boundaries of objects do not coincide with finite-difference grid lines, the staircase approximation is commonly used [2], [3]. However, the staircasing meshing can lead to inaccuracies electromagnetic simulations. Therefore, a shortcoming can be overcome by using conformal FDTD (CFDTD) techniques [4], [5]. Unfortunately, the above-mentioned CFDTD formulation can only handle cells filled with a homogenous dielectric; hence, it is not well suited for cells containing arbitrarily shaped dielectric objects with two or more dielectric materials. To solve this difficulty, the effective permittivities at dielectric-dielectric interface suitable for second-order accuracy have been obtained by applying of the numerical reflection coefficient [6]. And then, they were theoretically investigated in a rigorous application of the integral form of Maxwell's curl equations on the Yee's lattice leads to the desired values in a straightforward fashion [7].

In this paper, the effective permittivities of the dielectric interface are theoretically proved by using the electric properties of the boundary conditions.

Generally, in calculation of effective permittivities, the tangential component of electric field and the normal component of electric flux density are analyzed carefully because of the discontinuity of the fields at the dielectric interfaces.

II. EFFECTIVE PERMITTIVITIES

In the boundary conditions, it is proved that the tangential component of electric field is continuous across a boundary for which permittivity changes discontinuously. It is also seen that the normal component of the electric flux density is continuous across the discontinuity of electric boundary [8]. Typically, Fig. 1 shows one example of discontinuous distribution of permittivity across interface of two media where the relations between the field components can be obtained as equations (1) and (2) by the boundary conditions. The relations derived above are very useful in determining the discontinuous change of the fields at the boundary.

$$E_{y1}\cos\theta + E_{z1}\sin\theta = E_{y2}\cos\theta + E_{z2}\sin\theta \tag{1}$$

$$-\varepsilon_1 E_{v1} \sin \theta + \varepsilon_1 E_{z1} \cos \theta = -\varepsilon_2 E_{v2} \sin \theta + \varepsilon_2 E_{z2} \cos \theta \quad (2)$$

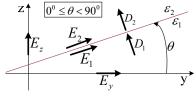


Fig. 1. The electric properties of boundary conditions

In the Fig. 2, schematic drawings of analyzed y-z and x-z planes for effective permittivity calculation are shown in detail. Let consider a boundary surface S separating a region with parameters $\varepsilon_1, E_{y1}, E_{z1}$ one side and $\varepsilon_2, E_{y2}, E_{z2}$ one the adjacent side. Here, ε_1 and ε_2 are the permittivities assigned to the region1 and region2, respectively. E_{y1}, E_{y2}, E_{z1} and E_{z2} are the electric fields with respect to the y- and z- directions, alternatively. The cell sizes are Δy

and Δz . The unknown effective permittivities ε_a and ε_b are assigned to E_y and E_z nodes nearest the dielectric interface. The effective permittivity ε_a associated with the electric component towards y- direction E_y at the dielectric interface is firstly considered. At the dielectric interface, in order to obtain expressions for the stored electric flux density D_y in the volume $V_a = \Delta x \cdot \Delta y \cdot \Delta z$, the volume integral of D_y over V_a is taken by using (3).

$$\int_{Va} \varepsilon_a \cdot E_y dV = \int_{Va1} \varepsilon_1 \cdot E_{y1} dV + \int_{Va2} \varepsilon_2 \cdot E_{y2} dV$$
 (3)

On the other hand, the electric flux in the volume V_a is obtained by using (4).

$$\int_{V_a} E_y dV = \int_{V_{a1}} E_{y1} dV + \int_{V_{a2}} E_{y2} dV$$
 (4)

From (3) and (4), the component of ε_a can be found as follows:

$$\varepsilon_{a} = \frac{\int_{V_{a}}^{\varepsilon_{a}} \varepsilon_{y} dV}{\int_{V_{a}}^{\varepsilon_{b}} E_{y} dV} = \frac{\int_{V_{a1}}^{\varepsilon_{1}} \varepsilon_{1} \varepsilon_{y1} dV + \int_{V_{a2}}^{\varepsilon_{2}} \varepsilon_{2} \varepsilon_{y2} dV}{\int_{V_{a1}}^{\varepsilon_{2}} E_{y1} dV + \int_{V_{a2}}^{\varepsilon_{2}} E_{y2} dV}$$
(5)

In a similar expression of ε_a calculation, the effective permittivity ε_b is given by (6).

$$\varepsilon_b = \frac{\int_{V_b} \varepsilon_b \cdot E_z dV}{\int_{V_b} E_z dV} = \frac{\int_{V_{b1}} \varepsilon_1 \cdot E_{z1} dV + \int_{V_{b2}} \varepsilon_2 \cdot E_{z2} dV}{\int_{V_{b1}} E_{z1} dV + \int_{V_{b2}} E_{z2} dV}$$
(6)

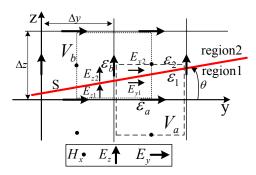
Equations (5) and (6) state that at the interface of dielectric media, the effective permittivities ε_a and ε_b are determined in the way to preserve the electric flux and the electric flux density limited by the corresponding volume.

From the expressions of (5) and (6), the relations between the effective permittivities and ratios of the electric fields at the boundary surface S are given by

$$\varepsilon_{a} = \frac{\varepsilon_{1} \cdot V_{a1} + \varepsilon_{2} \cdot V_{a2} (E_{y2} / E_{y1})}{V_{a1} + V_{a2} (E_{y2} / E_{y1})}$$
(7)

$$\varepsilon_b = \frac{\varepsilon_1 \cdot V_{b1} + \varepsilon_2 \cdot V_{b2} (E_{z2} / E_{z1})}{V_{b1} + V_{b2} (E_{z2} / E_{z1})}$$
(8)

where the ratios of the electric fields E_{y2}/E_{y1} and E_{z2}/E_{z1} can be found by the electric properties of the boundary condition in (1) and (2).



(a) y-z plane

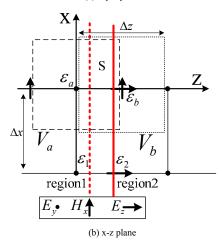


Fig. 2. Schematic drawings of analyzed y-z and x-z planes for the effective permittivity calculation

Since the angle θ made by the boundary surface S and the x-y plane of FDTD grid is zero, (1) and (2) now become

$$\frac{E_{y2}}{E_{y1}} = 1 \tag{9}$$

$$\frac{E_{z2}}{E_{z1}} = \frac{\varepsilon_1}{\varepsilon_2} \tag{10}$$

Substituting (9) and (10) into (7) and (8):

$$\varepsilon_a = \frac{\varepsilon_1 \cdot V_{a1} + \varepsilon_2 \cdot V_{a2}}{V_{a1} + V_{a2}} \tag{11}$$

$$\frac{1}{\varepsilon_b} = \frac{V_{b1}}{V_b} \frac{1}{\varepsilon_1} + \frac{V_{b2}}{V_b} \frac{1}{\varepsilon_2} \tag{12}$$

III. NUMERICAL EXPERIMENTS AND DISCUSSIONS

We can summarize the above calculated results of the effective permittivities by starting the numerical experiments. It is assumed that an electromagnetic wave is transmitted from dielectric medium 1 (ε_1) to dielectric medium 2 (ε_2) with an incident angle θ_i .

At the interface of two media where a mount of wave is reflected back to the dielectric medium 1 while the other part is transmitted into the dielectric medium 2, the numerical reflection coefficient is calculated by using (11) and (12). In this paper, we assume that the propagation constants of the numerical reflection coefficients are the ideal case. The incident wave is assigned by 374.92 MHz and the cell size is varied to check the accuracy of the effective permittivities.

On the other hand, the calculated numerical reflection coefficient is also compared with calculated results of the conventional method [9] and the theoretical analysis [1]. In [9], the effective permittivities is expressed as

$$\varepsilon_{eff} = \varepsilon_1 \cdot V(i, j, k) + \varepsilon_2 \cdot (1 - V(i, j, k)) \tag{13}$$

where V(i, j, k) is the fractional volume of the (i, j, k) th cell that contains the dielectric material ε_1 , and ε_2 is the dielectric constant of the medium in the remaining volume. In order to evaluate the calculated accuracy of the effective permittivities while the boundary surface S is orthogonal to the FDTD grid, an example for the effective permittivity calculation is presented by Fig. 3.

By following the proposed method in (11) and (12), we obtain

$$\varepsilon_a = (1/2 + d)\varepsilon_1 + (1/2 - d) \cdot \varepsilon_2 \tag{14}$$

$$\frac{1}{\varepsilon_b} = \frac{d}{\varepsilon_1} + \frac{1 - d}{\varepsilon_2} \tag{15}$$

where d is the offset ratio satisfied the condition $0 \le d \le 1/2$. The permittivities ε_1 and ε_2 are assigned to the region1 and the region2, respectively.

By employing the solutions of (14) and (15), the effective permittivities ε_a and ε_b are used to compute the numerical

reflection coefficients and the calculated results are compared with that of the conventional method and the exact result of the theoretical analysis, as shown in Fig. 4. In here, the numerical reflection coefficient is considered in the change of cell size in x-, y-, z- direction, respectively.

From the numerical experiments in Fig. 4, it is easily shown that the calculated reflection coefficients are good agreement with the theoretical calculation in variation of incident angle. The comparisons show the proposed method as an exact method in the calculation of the effective dielectric constants with high accuracy.

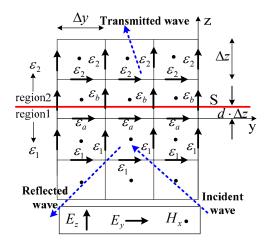
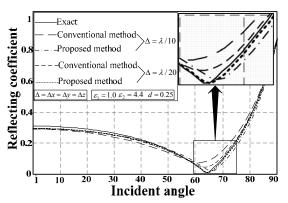
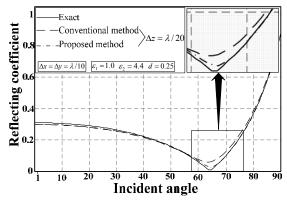


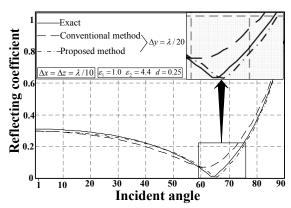
Fig. 3. Example for the effective permittivity calculation



(a) The same variation of the cell sizes in x-, y-, z- directions



(b) The variation of the cell size in z- direction



(c) The variation of the cell size in y- direction

Fig. 4. Illustration for comparison of reflection coefficients

IV. CONCLUSIONS In this paper, the solutions of effective permittivities at the

dielectric interface in the Yee's FDTD algorithm are theoretically investigated by using the boundary conditions. On the other hand, the accuracy of analyzed results is expressed by the numerical experiments employing the numerical reflection coefficients. Therefore, the consideration of the maintaining of the electric properties in the computation of the effective permittivities leads us to conclude that at the interface of two dielectric media, the boundary conditions as well as the characteristics of wave impedance and wave reflection coefficient are prevented from unphysical phenomena.

The numerical experiments show that the proposed method for the effective permittivity calculation is an exact solution with high accuracy. The good performances of the numerical experiments above confirm that.

ACKNOWLEDGEMENT

This research was supported by the Program for the Training of Graduate Students in Regional Innovation which was conducted by the Ministry of Commerce, Industry and Energy of the Korea Government.

REFERENCES

- [1] A. Taflove, Computational Electrodynamics, the Finite-Difference Time-Domain Method. Norwood, MA: Artech House, 1995.
- [2] Andreas C. Cangellaris and Diana B. Wright, "Analysis of the Numerical Error Caused by the Stair-Stepped Approximation of a Conducting Boundary in FDTD Simulations of Electromagnetic Phenomena," *IEEE Transactions on Antennas and Propagation*, vol. 39, no. 10, October 1991.
- [3] A. Akyurtlu, D. H. Werner, V. Veremey, D. J. Steich and K. Aydin, "Staircasing Errors in FDTD at an Air– Dielectric Interface," *IEEE Microwave and Guided Wave Letters*, vol. 9, no. 11, November 1999.
- [4] W. Yu and R. Mittra, "A Conformal Finite Difference Time Domain Technique for Modeling Curved Dielectric Surfaces," *IEEE Microwave and Wrireless Components Letter*, vol. 11, no. 1, January 2001.
- [5] G. Waldschmidt and A. Taflove, "Three-Dimensional CAD-Based Mesh Generator for the Dey–Mittra Conformal FDTD Algorithm," *IEEE Transactions on Antennas and Propagation*, vol. 52, no. 7, July 2004.
- [6] T. Hirono, Y. Shibata, W. W. Lui, S. Seki, and Y. Yoshikuni, "The second-order condition for the dielectric interface orthogonal to the Yee-lattice axis in the FDTD scheme," *IEEE Microwave Guided Wave Letters*, vol. 10, pp. 359–361, September 2000.
- [7] K. Hwang and A. C. Cangellaris, "Effective Permittivities for Second-Order Accurate FDTD Equations at Dielectric Interfaces," *IEEE Microwave and Wireless Components Letters*, vol. 11, no. 4, April 2001.
- [8] R. E.Collin, Field Theory of Guided Waves. IEEE PRESS, 1990.
- [9] S. Dey and R. Mittra, "A conformal finite-difference time-domain technique for modeling cylindrical dielectric resonators," *IEEE Microwave Theory and Techniques*, vol. 47, no. 9, pp. 1737–1739, September 1999.