

STUDY OF INPUT IMPEDANCE OF RECTANGULAR MICROSTRIP ANTENNAS

Zhou Tai Zhang Zhichao

(Dept. of Electronic Engineering , Shandong University ,Jinan,P.R.China)

ABSTRACT

This paper gives a new model of current density in the patch of rectangular microstrip antennas . Using improved moment method with fewer terms of basis functions , formulas for calculating the impedance are derived and software is achieved .The results show that higher precision can be obtained without any correction .

Key words: microstrip antennas, current density model, moment method, input impedance

1 Instruction

After proposed in 1953 , microstrip antennas have been developed rapidly until 1970's . With its special advantages , microstrip antennas are used widely in space technology and earth equipment in recent years [1,2].

The input impedance is a very important term in the design and analysis of microstrip antennas. Among the analysing methods , integral equation method has special advantages over transmission line method and cavity model theory method [1-4], and if it is connected with moment method , it can be used widely .

Using integral equation and moment method with fewer terms of basis function , in order to achieve useful results in wider frequency , this paper gives a new model of the current density on the patch and then studies the input impedance of rectangular microstrip patch antenna fed by a probe .

2 The formulas for input impedance

To solve the input impedance of microstrip antennas with moment method

, use the reactive intergral equation :

$$\int_s \vec{J}_s \cdot \vec{E}_T dS = - \int_{V_i} \vec{J}_i \cdot \vec{E}_T dV \quad (1)$$

Give the current density expression on the patch as:

$$\vec{J}_s(X,Y) = \sum_{n=1}^N I_n \cdot \vec{J}_n(x,y) + \vec{J}_0 \quad (2)$$

in it :

$$\vec{J}_0 = \vec{x}J_{0x} + \vec{y}J_{0y}$$

and :

$$J_{0x} = \begin{cases} -X/L-1/2 & -L/2 < X < X_s \\ -X/L+1/2 & X_s < X < L/2 \end{cases}$$

$$J_{0y} = \begin{cases} -Y/W-1/2 & W/2 < Y < Y_s \\ -Y/W+1/2 & Y_s < Y < W/2 \end{cases}$$

it can be obtained easily :

$$\vec{J}_s(Kx,Ky) = \sum_{n=1}^N I_n \cdot \vec{J}_n(Kx,Ky) + \vec{J}_0(Kx,Ky)$$

and :

$$\vec{E}_T(x,y) = \frac{1}{4\pi^2} \iint_{-\infty}^{\infty} \vec{E}_T(Kx,Ky) \cdot e^{-j(KxX+KyY)} dKx dKy \quad (3)$$

$$\vec{E}_T(Kx,Ky) = \vec{G}(Kx,Ky) \cdot \vec{J}_s(Kx,Ky) \quad (4)$$

where : $\vec{G}(Kx,Ky)$ is the dyadic Green's function in the space .

Substitute formula (2) and (3) into (1) , and then substitute (4) into (1) , we can obtain :

$$[Z_{nm}][I_n] = [V_n]$$

And then we can solve out $[I_n]$, after that , we can obtain $\vec{J}_s(Kx,Ky)$,

$\vec{E}_T(Kx,Ky)$ and so on:

$$P_{in} = -\frac{1}{8\pi^2} \iint_{-\infty}^{\infty} \vec{E}_T(Kx,Ky) \cdot \vec{J}_s(Kx,Ky) dKx dKy \quad (5)$$

but :

$$P_{in} = I_0^2 \cdot Z_{in}$$

and so :

$$Z_{in} = P_{in} / I_0^2 \quad (6)$$

3 Software and some expression [5]

To obtain calculation of input impedance, we use an adaptive automatic integration based on Newton-Cotes 9 point rule. Adaptive automatic integration is a method which automatically places abscissas density where the integrand $f(x)$ changes rapidly, or sparsely where it changes gradually. This is currently a very good method for automatic integration in terms of reliability and economy.

In the procedure of solving the input impedance, many variations are complex. To avoid complicated complex calculation, we do the integration for the real part and imaginary part separately.

In order to achieve more exactly results, we at first solve the singularities in the integrand, and then take the singularities as endpoint, to divide integration interval into some subintervals with the singularities as endpoints, and then do the integrations at each subintervals.

This software is achieved with FORTRAN language and passed on the computer MS-340.

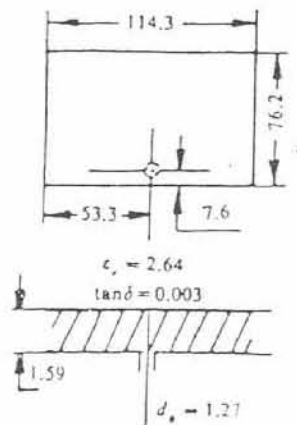


Fig.1 Rectangular microstrip patch antenna

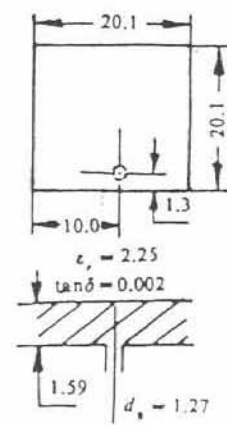


Fig.2 Rectangular microstrip patch antenna

4 Results.

Using improved moment method, we give the new model of the current density on the patch. And with two terms of basis function (except \vec{J}_0):

$$\begin{aligned} \vec{J}_s(x,y) &= \vec{x}\vec{I}_1 \cdot J_1(x,y) + \vec{y}\vec{I}_2 \cdot J_2(x,y) + \vec{J}_0 \\ &= \vec{x}\vec{I}_1 \cdot \sin(\pi X/L) + \vec{y}\vec{I}_2 \cdot \sin(\pi Y/W) + \vec{J}_0 \end{aligned}$$

By calculating the input impedance of two examples (sizes show as Fig.1 and Fig.2), the results agree very well with that of experiments and calculation in [3], show as in Fig.3 and Fig.4. But the method doesn't need the step to correct calculated results as [3], so this method has great values for the studies of microstrip antennas with other shape of patch ,or with many substrate layers covered on it .

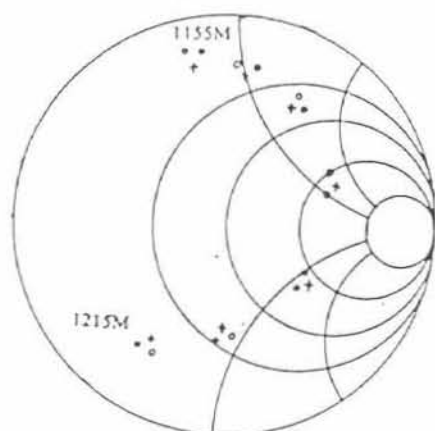


Fig.3 Input impedance of rectangular microstrip patch antenna

(showed in Fig.1)

- Calculated [3]
 - Measured [3]
 - + Calculated this paper
- Increment :10 MHZ
(Increasing clockwise)

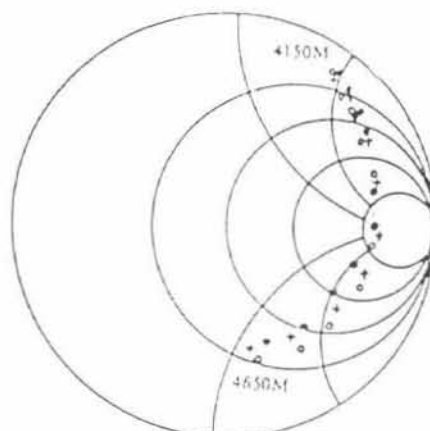


Fig.4 Input impedance of rectangular microstrip patch antenna

(showed in Fig.2)

- Calculated [3]
 - Measured [3]
 - + Calculated this paper
- Increment: 50 MHZ
(Increasing clockwise)

REFERENCES

- 1 I J Ball and P Bhartia Microstrip Antennas . Artech House , 1980
- 2 J R James , P S Hall and C Wood ,Microstrop Antennas Theory and Design . Peter Perginus,Ltd ,1981
- 3 M D Deshparde and M C Baily. IEEE Trans ,1982,AP-30:645-650
- 4 Y T Lo,D Solomon and W F Richards . IEEE Trans ,1979,AP-27:137-145
- 5 CURTIS F GERALD . Applied Numerical Analysis . Addision-Wesley Publishing Company .