THE FOCUSED FIELDS OF FRESNEL ZONE PLATE LENS*

Guo Zhong JIANG Wen Xun ZHANG Kai KANG State Key Lab. of Millimeter Waves, Southeast University Naniing 210096 CHINA

INTRODUCTION

Because of its low cost, light weight, and easy fabrication, the Fresnel Zone Plate (FZP) Antenna is becoming an important candidate for such applications as DBS reception and mobile communication. Since the fact of feed blockage in FZP reflector [1], the study of FZP lens is excited. In this paper a dual-layer FZP lens (Fig. 1) is proposed to enhance the aperture efficiency, the field intensity in focal region under plane wave normal incidence is studied, and compared with single-layer FZP lens. An improved full-wave analysis based on vector Hankel transform and spectral domain immitance matrix approach is developed, with several attractive features: 1) the Green's functions in spectral domain can be derived by using equivalent transmission line method; 2) only single integration are required; 3) only low-order matrix is to be dealt with. Numerical results match the physical characteristics of FZP lens.

ANALYSIS

The transverse components of electromagnetic fields $(E_{\rho}, E_{\varphi}, H_{\rho}, H_{\varphi})$ in FZP structure (Fig.1) holding hybrid modes $(TM^z + TE^z)$ can be formulated by both electric and magnetic Hertz potentials π^e and π^h in \hat{z} direction with time-harmonic factor $e^{j\omega t}$ and periodic factor $e^{jn\varphi}$ (both are omitted below) as follows:

$$\mathbf{E}_{\perp n} = \begin{bmatrix} E_{\rho n}(\rho, z) \\ E_{\varphi n}(\rho, z) \end{bmatrix} = \begin{bmatrix} \frac{\partial}{\partial \rho} \frac{\partial}{\partial z} & \omega \mu \frac{n}{\rho} \\ \frac{jn}{\rho} \frac{\partial}{\partial z} & j\omega \mu \frac{\partial}{\partial \rho} \end{bmatrix} \begin{bmatrix} \pi_n^e(\rho, z) \\ \pi_n^h(\rho, z) \end{bmatrix}, \quad \mathbf{H}_{\perp n} = \begin{bmatrix} H_{\rho n}(\rho, z) \\ H_{\varphi n}(\rho, z) \end{bmatrix} = \begin{bmatrix} \frac{\partial}{\partial \rho} \frac{\partial}{\partial z} & -\omega \varepsilon \frac{n}{\rho} \\ \frac{jn}{\rho} \frac{\partial}{\partial z} & -j\omega \varepsilon \frac{\partial}{\partial \rho} \end{bmatrix} \begin{bmatrix} \pi_n^h(\rho, z) \\ \pi_n^e(\rho, z) \end{bmatrix}$$
(1)

Introducing the vector Hankel transform pair [3]:

$$\widetilde{\mathbf{f}}_{n}(k_{\rho}) = \int_{0}^{\infty} \mathcal{A}_{n}(k_{\rho}\rho) \mathbf{f}_{n}(\rho) \rho d\rho , \qquad \qquad \mathbf{f}_{n}(\rho) = \int_{0}^{\infty} \mathcal{A}_{n}(k_{\rho}\rho) \widetilde{\mathbf{f}}_{n}(k_{\rho}) k_{\rho} dk_{\rho} \tag{2}$$
where
$$\widetilde{\mathbf{f}}_{n}(k_{\rho}) = \begin{bmatrix} \widetilde{f}_{1n}(k_{\rho}) \\ \widetilde{f}_{2n}(k_{\rho}) \end{bmatrix}, \quad \mathbf{f}_{n}(\rho) = \begin{bmatrix} f_{1n}(\rho) \\ f_{2n}(\rho) \end{bmatrix}, \quad \mathcal{A}_{n}(k_{\rho}\rho) = \begin{bmatrix} J'_{n}(k_{\rho}\rho) & -j\frac{n}{k_{\rho}\rho} J_{n}(k_{\rho}\rho) \\ j\frac{n}{k_{\rho}\rho} J_{n}(k_{\rho}\rho) & J'_{n}(k_{\rho}\rho) \end{bmatrix}$$

 $J_n(\bullet)$ and $J_n'(\bullet)$ are the Bessel function and its derivative, respectively, the fields $\widetilde{\mathbf{E}}_{1n}$ and $\widetilde{\mathbf{H}}_{1n}$ in spectral domain may be expressed as:

$$\widetilde{\mathbf{E}}_{\perp n} = \begin{bmatrix} \widetilde{E}_{\rho n} \\ \widetilde{E}_{\varphi n} \end{bmatrix} = \int_{0}^{\infty} \rho \, d\rho \begin{bmatrix} \mathcal{L}_{l} & 0 \\ 0 & \mathcal{L}_{2} \end{bmatrix} \begin{bmatrix} \pi^{e} \\ \pi^{h} \end{bmatrix}, \quad \widetilde{\mathbf{H}}_{\perp n} = \begin{bmatrix} \widetilde{H}_{\rho n} \\ \widetilde{H}_{\varphi n} \end{bmatrix} = \int_{0}^{\infty} \rho \, d\rho \begin{bmatrix} 0 & \mathcal{L}_{3} \\ \mathcal{L}_{4} & 0 \end{bmatrix} \begin{bmatrix} \pi^{e} \\ \pi^{h} \end{bmatrix}$$
(3)

^{*}This work is supported by the National Natural Science Fund (No.69571009) of China

with the operators:

$$\mathcal{L}_1 = \mathcal{L}_3 = \left[J_n' \left(k_\rho \rho \right) \frac{\partial}{\partial \bar{\rho}} + \frac{n^2}{k_\rho \rho^2} J_n \left(k_\rho \rho \right) \right] \frac{\partial}{\partial z} , \quad \mathcal{L}_2 = -\frac{\varepsilon}{\bar{\mu}} \mathcal{L}_4 = j \omega \mu \left[J_n' \left(k_\rho \rho \right) \frac{\partial}{\partial \rho} + \frac{n^2}{k_\rho \rho^2} J_n \left(k_\rho \rho \right) \right]$$

Thus the hybrid modes in spatial domain are decomposed into independent TM modes $(\tilde{E}_{zn}, \tilde{E}_{\rho n}, \tilde{H}_{\varphi n})$ due to π^e and TE modes $(\tilde{H}_{zn}, \tilde{E}_{\varphi n}, \tilde{H}_{\rho n})$ due to π^h in spectral domain.

Furthermore, the vector Hankel transform of the magnetic-field boundary condition on the patches from spatial to spectral domain are:

$$\begin{bmatrix} \widetilde{H}_{\rho n1} \\ \widetilde{H}_{\omega n1} \end{bmatrix} - \begin{bmatrix} \widetilde{H}_{\rho n2} \\ \widetilde{H}_{\omega n2} \end{bmatrix} = \begin{bmatrix} \widetilde{I}_{\rho n} \\ \widetilde{I}_{\omega n} \end{bmatrix} = \int_{0}^{\infty} \mathcal{A}(k_{\rho}\rho) \begin{bmatrix} I_{\rho n} \\ I_{\varphi n} \end{bmatrix} \rho \, d\rho \tag{4}$$

where $\tilde{I}_{\rho n}$ and $\tilde{I}_{\sigma n}$ correspond to the TM and TE modes, respectively.

By means of the viewpoint of equivalent transmission lines. (Fig.2), the propagation constant is $\gamma^2 = k_\rho^2 - k_0^2$, the wave admittances are $Y_c^e = j\omega\varepsilon_0/\gamma$, $Y_c^h = \gamma/j\omega\mu_0$. The Green's function, i.e. the electric fields produced by unit current source may be understood as impedance elements in the matrix equation:

$$\begin{bmatrix} \widetilde{\mathbf{E}}_{\perp n_1} \\ \widetilde{\mathbf{E}}_{\perp n_2} \end{bmatrix} = \begin{bmatrix} \widetilde{z}_{11} \\ \widetilde{z}_{21} \end{bmatrix} \begin{bmatrix} \widetilde{z}_{12} \\ \widetilde{z}_{22} \end{bmatrix} \begin{bmatrix} \widetilde{\mathbf{I}}_{n_1} \\ \widetilde{\mathbf{I}}_{n_2} \end{bmatrix}$$
(5)

where
$$\begin{bmatrix} \widetilde{\mathbf{E}}_{\perp ni} \end{bmatrix} = \begin{bmatrix} \widetilde{E}_{\rho ni} \\ \widetilde{E}_{\varphi ni} \end{bmatrix}$$
, $\begin{bmatrix} \widetilde{\mathbf{I}}_{\perp ni} \end{bmatrix} = \begin{bmatrix} \widetilde{I}_{\rho ni} \\ \widetilde{I}_{\varphi ni} \end{bmatrix}$, $\begin{bmatrix} \widetilde{z}_{ij} \end{bmatrix} = \begin{bmatrix} \widetilde{z}_{ij}^e & 0 \\ 0 & \widetilde{z}_{ij}^h \end{bmatrix}$, $\widetilde{z}_{ij}^{e,h} = e^{-\gamma |z_j - z_i|} / 2Y_c^{e,h}$ $(i, j = 1, 2)$.

Substitute (5) into electric-field boundary condition on the patches, a couple of integral equations of $\widetilde{\mathbf{I}}_{ni}$ is modelled as:

$$\begin{bmatrix} E_{\rho ni} \\ E_{\varphi ni} \end{bmatrix} = \int_0^\infty \mathcal{A}_n \left(k_\rho \rho \right) \begin{bmatrix} [z_{11}] & [z_{12}] \\ [z_{21}] & [z_{22}] \end{bmatrix} \begin{bmatrix} \widetilde{I}_{\rho ni} \\ \widetilde{I}_{\varphi ni} \end{bmatrix} k_\rho dk_\rho = - \begin{bmatrix} E_{\rho ni}^{inc} \\ E_{\alpha ni}^{inc} \end{bmatrix}$$
(6)

By the Galerkin's procedure, the spectral current in (6) is expanded as:

$$\widetilde{\mathbf{I}}_{nl} = \begin{bmatrix} \widetilde{I}_{\rho nl} \\ \widetilde{I}_{\phi nl} \end{bmatrix} = \sum_{l=1}^{L} \left(\sum_{m=1}^{M} \alpha_{nilm} \widetilde{\mathbf{e}}_{nilm} + \sum_{n=1}^{P} \beta_{nilp} \widetilde{\mathbf{h}}_{nilp} \right)$$
(7)

where the basis functions $\{\tilde{\mathbf{e}}_{nilm}, \tilde{\mathbf{h}}_{nilp}\}$ transformed from the eigen-functions $\{\mathbf{e}_{nilm}, \mathbf{h}_{nilp}\}$ defined on the rings bounded by $r_{il+} = \sqrt{2l\lambda F_i/v + (l\lambda/v)^2}$ and $r_{il-} = \sqrt{(2l-1)\lambda F_i/v + (2l-1/2v)^2}$ in spatial domain for TM and TE modes[2] respectively, where i = 1,2 means the number of layer; $l = 1,2,\cdots$ means the number of Fresnel ring in each plate, and F_i is the i-th focal of FZP.

$$\widetilde{\mathbf{e}}_{nilm} = \begin{bmatrix}
\frac{-x'_{nilm}}{k_{\rho}^{2} - x'_{nilm}^{2}} \left(r_{il+}J_{n}'(k_{\rho}r_{il+})\psi_{nilm}^{e}(r_{il+}) - r_{il-}J_{n}'(k_{\rho}r_{il-})\psi_{nilm}^{e}(r_{il-})\right) \\
j \frac{n}{k_{\rho}x'_{nilm}} \left(J_{n}(k_{\rho}r_{il+})\psi_{nilm}^{e}(r_{il+}) - J_{n}(k_{\rho}r_{il-})\psi_{nilm}^{e}(r_{il-})\right) \\
0 \\
j \frac{k_{\rho}}{k_{\rho}^{2} - x_{nilp}^{2}} \left(r_{il+}J_{n}(k_{\rho}r_{il+})\psi_{nilp}^{h}(r_{il+}) - r_{il-}J_{n}(k_{\rho}r_{il-})\psi_{nilm}^{h'}(r_{il-})\right)
\end{bmatrix} (8)$$

where

$$\psi_{nilm}^{e}(\rho) = N'_{n}(x'_{nilm}r_{il-})J_{n}(x'_{nilm}\rho) - J'_{n}(x'_{nilm}r_{il-})N_{n}(x'_{nilm}\rho)$$

$$\psi_{nilp}^{h}(\rho) = N_{n}(x_{nimp}r_{il-})J_{n}(x_{nilp}\rho) - J_{n}(x_{nilp}r_{il-})N_{n}(x_{nilp}\rho)$$

 x'_{nilm} is the m-th root of $\psi_{nilp}^{e'}$, x_{nilp} is the p-th root of $\psi_{nilp}^{h'}$, the Hermitain inner-products with $\left\{\rho\widetilde{\mathbf{e}}_{nilm}, \rho\widetilde{\mathbf{h}}_{nilp}\right\}$ are taken for equation (6), also the Parseval's theorem for vector Hankel transform should be employed. Finally a set of simultaneous algebraic equations is formulated in the matrix form as:

$$\begin{bmatrix} A_{11}^{ee} & A_{12}^{ee} & A_{11}^{ee} \\ A_{21}^{ee} & A_{22}^{ee} & A_{21}^{eh} & A_{22}^{eh} \\ A_{21}^{he} & A_{12}^{he} & A_{11}^{hh} & A_{12}^{hh} \\ A_{21}^{he} & A_{22}^{he} & A_{21}^{hh} & A_{22}^{hh} \end{bmatrix} \begin{bmatrix} [\alpha_1] \\ [\alpha_2] \\ [\beta_1] \end{bmatrix} = \begin{bmatrix} F_1^e \\ F_2^e \\ F_1^h \\ F_2^h \end{bmatrix}$$

$$\begin{bmatrix} A_{11}^{he} & A_{12}^{he} \\ A_{21}^{he} & A_{22}^{he} \\ A_{21}^{he} & A_{22}^{hh} \end{bmatrix} \begin{bmatrix} [\alpha_1] \\ [\alpha_2] \\ [\beta_2] \end{bmatrix} = \begin{bmatrix} F_1^e \\ F_2^h \\ F_2^h \end{bmatrix}$$

$$(9)$$

where:
$$A_{ij}^{ee} = \int_{0}^{\infty} k_{\rho} dk_{\rho} \widetilde{\mathbf{e}}_{nil'nl}^{*} \left[\widetilde{z}_{ij} \right] \widetilde{\mathbf{e}}_{njlm}, \quad A_{ij}^{eh} = \int_{0}^{\infty} k_{\rho} dk_{\rho} \widetilde{\mathbf{e}}_{nil'nl}^{*} \left[\widetilde{z}_{ij} \right] \widetilde{\mathbf{h}}_{njlp}, \quad F_{i}^{e} = \int_{r_{il'}}^{r_{il'}} \left(\mathbf{e}_{nil'nl'}^{*} \cdot \mathbf{E}_{n}^{inc} \right) \rho d\rho$$

$$A_{ij}^{he} = \int_{0}^{\infty} k_{\rho} dk_{\rho} \widetilde{\mathbf{h}}_{nil'p'}^{*} \left[\widetilde{z}_{ij} \right] \widetilde{\mathbf{e}}_{njlm}, \quad A_{ij}^{lh} = \int_{0}^{\infty} k_{\rho} dk_{\rho} \widetilde{\mathbf{h}}_{nil'p'}^{*} \left[\widetilde{z}_{ij} \right] \widetilde{\mathbf{h}}_{nylp}, \quad F_{i}^{h} = \int_{r_{il'}}^{r_{il'}} \left(\mathbf{h}_{nil'p'}^{*} \cdot \mathbf{E}_{n}^{inc} \right) \rho d\rho$$

The current expansion coefficients (α_i, β_i) can be solved from (9). By using equivalent transmission lines (Fig.2), we get the spectral scattered electric field components in the focal region of a dual-layer FZP lens as follows:

$$\begin{bmatrix} \widetilde{E}_{\rho n}^{sc} \\ \widetilde{E}_{\rho n}^{sc} \end{bmatrix} = \sum_{k=1}^{2} \sum_{l=1}^{L} \begin{bmatrix} \widetilde{E}_{\rho nilm}^{sc}(z) \\ \widetilde{E}_{\rho nilm}^{sc}(z) \end{bmatrix}, \begin{bmatrix} \widetilde{E}_{\rho nilm}^{sc}(z) \\ \widetilde{E}_{\rho nilm}^{sc}(z) \end{bmatrix} = \frac{e^{-\gamma|z-z_i|}}{2} \begin{bmatrix} 1/\gamma_c^{c} & 0 \\ 0 & 1/\gamma_c^{h} \end{bmatrix} \widetilde{J}_{\rho i}$$

$$(10)$$

Since hybrid modes in spectral domain have been decomposed into independent *TM* and *TE* modes, the spectral scattered magnetic components can be got as:

$$\widetilde{H}_{\varphi n}^{sc} = -Y_c^v \widetilde{E}_{\rho n}^{sc}, \quad \widetilde{H}_{\rho n}^{sc} = Y_c^h \widetilde{E}_{\varphi n}^{sc}$$
(11)

Once the transverse components of scattered fields in spatial domain $E_{\rho n}^{sc}$, $E_{\rho n}^{sc}$, $H_{\rho n}^{sc}$, $H_{\rho n}^{sc}$, $H_{\rho n}^{sc}$ are solved by means of the inverse vetor Hankel transform of their spectral expressions, the z components E_{zn}^{sc} , H_{zn}^{sc} can also be got according to Maxwell's equations.

NUMERICAL RESULTS

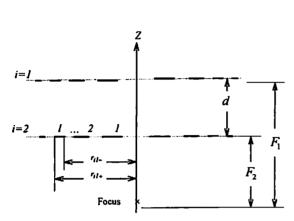
For a FZP with λ/ν phase complement[1], a half-wave ($\nu=2$) dual-layer FZP lens with the focal lengths $F_1=0.20m$, $F_2=F_1-d$, is designed at f=9.375GHz, each layer consists of four (l=1,2,3,4) rings. Denote $R=10\lg\left(\left|E^{sc}\right|^2/\left|E^{inc}\right|^2\right)$ is scattered field level in focal region, for single-layer (R_1) or dual-layer (R_2) FZP lens, the difference (R_2-R_1) is a function of d, which shows the gain enhancement due to the second FZP (Fig.3); and that $d=\lambda/2$ is a best choice, while the field distributions around the focus are shown in Fig.4 and Fig.5, respectively. Obviously, the field intensity is concentrated at focus with a region of 3dB reduction about $2\lambda \times 12\lambda$ along and across the focal axis. The results of field distribution are matching the physical characteristics of Fresnel Zone Plate Antenna.

REFERENCES

[1]J.C. Wiltse and J.E. Garret, The Fresnel Zone Plate Antenna, Microwave Journal, Jan., 1991

[2]S.M.Ali, W.C.Chew, Vector Hankel Transform Analysis of Annular-Ring Microstrip Antenna, IEEE, Trans., Vol. AP-30, No.4, July, 1982

[3]W.C., Chew, T.M. Habashy, the use of Vector Transforms in solving some Electromagnetic Scattering Problems, IEEE Trans., Vol. AP-34, No.7, July., 1986



2.0-1.5-0.0-

Fig.3. Gain enhancement of the 2nd FZP

Fig.1 Geometry of dual-layer FZP lens

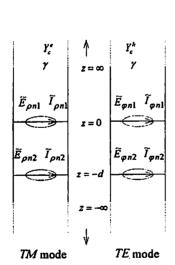


Fig.2 Equivalent circuit of dual-layer FZP lens in spectral domain

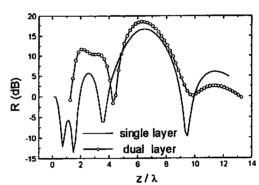


Fig.4 Field intensity along focal axis (d=3/2)

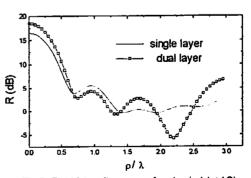


Fig.5 Field intensity across focal axis (d=3/2)