# An Implementation of Optimum-Time Firing Squad Synchronization Algorithm on 1-Bit Cellular Automaton 

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#### Abstract

We propose a time-optimum ( $2 n-2$ )step firing squad synchronization algorithm for a new class of cellular automata, CA1-bit, whose inter-cell communication amount at one step is restricted to 1 bit. The number of internal states of each cell implemented is 78 and the total number of transition rules is 208 .


## 1. Introduction

In the long history of the study of the CA, generally speaking, the number of internal states of each cell is finite and the local state transition rule is defined in a such way that the state of each cell depends on the previous states of itself and its neighboring cells. Thus, in the finite state description of the CA, the amount of communication bits exchanged at one step between neighboring cells is assumed to be $\mathrm{O}(1)$ bit, however, such bit-information exchanged between inter-cell has been hidden behind the definition of conventional automata-theoretic finite state description.

In the present paper, we focus our attention to the communication bits exchanged between inter-cells and introduce a new class of cellular automata, CA1-bit, whose inter-cell communication amount at one step is restricted to 1-bit. We refer the model as 1-bit CA. The number of internal states of the CA1-bit is assumed to be finite in a usual sense. The next state of each cell is determined by the present state of itself and two binary 1-bit inputs from its left and right neighbor cells. Thus the 1-bit CA can be thought of as one of the most powerless and the simplest models in a variety of CAs. We study a classical firing squad synchronization problem that gives a finite-state protocol for synchronizing a large scale of cellular automata, in which it was originally proposed by J. Myhill to synchronize all parts of self-reproducing cellular automata [5]. The firing squad synchronization problem has been studied extensively for more than 40 years [1-11]. It is important and interesting to develop optimum-time algorithms on the most powerless and the simplest models in a variety of CAs. First, we introduce a cellular automaton with 1-bit inter-cell communication and define the firing squad synchronization problem on CA1-bit. Then,


Figure 1: A one-dimensional cellular automaton having 1-bit inter-cell communication.
we give an optimum-time firing squad synchronization algorithm on CA1-bit. The algorithm is based on the classical synchronization scheme developed by Waksman [11], having $\mathrm{O}(1)$-bit communication, and it will be implemented on a CA1-bit with 78 internal states and 208 transition rules.

## 2. A Firing Squad Synchronization Problem on $\mathrm{CA}_{1 \text {-bit }}$

### 2.1. 1-Bit Communication Cellular Automaton

A one-dimensional 1-bit inter-cell communication cellular automaton consists of an infinite array of identical finite state automata, each located at positive integer point. See Fig. 1. A cell at point $i$ is denoted by $\mathrm{C}_{i}$, where $1 \leq i \leq n$. Each $\mathrm{C}_{i}$, except $\mathrm{C}_{1}$ and $\mathrm{C}_{n}$, is connected with its left and right neighbor cells via a left or right one-way communication link, in which those communication links are indicated by right- and left-going arrows, as is shown in Fig. 1, respectively. Each one-way communication link can transmit only one bit at each step in each direction. The array operates in lock-step mode in such a way that the next state of each cell (except both end cells) is determined by both its own present state and the present binary inputs of its right and left neighbors. A more formal treatment can be found in Nishimura, Sogabe and Umeo $[6,7]$. The CA1-bit is a special subclass of normal (i.e., conventional) cellular automata studied so far.

### 2.2. Firing Squad Synchronization Problem on $\mathrm{CA}_{1-\mathrm{bit}}$

The firing squad synchronization problem is formalized in terms of the model of $\mathrm{CA}_{1-\text { bit }}$. All cells (soldiers), except the left and right end cells, are initially in the quiescent state at time $t=0$ with the property
that the next state of a quiescent cell with quiescent neighbors is the quiescent state again. At time $t=0$ the general cell $\mathrm{C}_{1}$ is in fire-when-ready state that is an initiation signal to the array. The firing squad synchronization problem $[1,5,11]$ is stated as follows: Given an array of $n$ identical cellular automata, including a general on the left end cell which is activated at time $t=0$, we want to give the description (state set and next-state function) of the automata so that, at some future time, all the cells will simultaneously and, for the first time, enter a special firing state. The set of states and the next-state function must be independent of $n$. The tricky part of the problem is that the same kind of soldier with a fixed number of states is required to synchronize, regardless of the length $n$ of the array.

## 3. Waksman's Optimum-Time Algorithm

### 3.1. Outline of Waksman's Algorithm

Waksman's algorithm is constructed on the conventional $\mathrm{O}(1)$-bit communication CA, and it can synchronize any cellular array consisting of $n$ cells at exactly $2 n-2$ steps. Figure 2 shows its timespace diagram for the synchronization. At time $t=0$, a general $G_{0}$, located at $\mathrm{C}_{1}$, generates an a signal and $k-1 \mathrm{~b}_{k^{-}}$ signals, where $2 \leq$ $k \leq\left\lceil\log _{2} n\right\rceil-1$. The a-signal propa-


Figure 2: Time-space diagram for Waksman's synchronization algorithm. gates right at the sloop of $\frac{1}{1}$ (one cell per step). The $b_{k}$-signal propagates in the right direction at the sloop of $\frac{1}{2^{k}-1}$ (one cell per $2^{k}-1$ step). The a-signal reaches $\mathrm{C}_{n}$ at $t=n-1$, and generates $\mathrm{G}_{1}$ there and then it reflects and proceeds in the left direction at the same speed. The reflected signal meets $\mathrm{b}_{2}$-signal, $\mathrm{b}_{3}$-signal, ..., $\mathrm{b}_{k}$-signal, and generates $k-1$ generals $\mathrm{G}_{2}, \mathrm{G}_{3}, \ldots, \mathrm{G}_{k}$ at each crossing point, respectively. Let $S$ be any cellular space and $|S|$ be size of $S$. Let $\mathrm{S}_{0}$ be the initial cellular space consisting of $n$ cells and $\mathrm{S}_{i}$ be a cellular space betwen $\mathrm{G}_{i+1}$ and $\mathrm{G}_{i}$. The general $\mathrm{G}_{i}$ is responsible for synchronizing $\mathrm{S}_{i}$ for any $i$ such that $1 \leq i \leq k$.

### 3.2. Generation of Generals

$\mathrm{G}_{i}$ is generated at the position which divides exactly $\left|S_{i-2}\right|$. Therefore the parity of $\left|S_{i-2}\right|$ is an important factor for generating $G_{i}$. If the parity of $\left|S_{i-2}\right|$ is an odd number, $\mathrm{G}_{i}$ consists of a cell. And if the parity of $\left|S_{i-2}\right|$ is an even number, $\mathrm{G}_{i}$ consists of two cells. This parity information is determined by the cell on $\mathrm{G}_{i-1}$, and it is transmitted to the cell of $\mathrm{G}_{i}$ by the a-signal. The signal can mark the general state $\mathrm{G}_{i}$ on the cell. Figure 2 is a time-space diagram for generating $G_{2}$ in case of that $n$ is an even number. $\mathrm{G}_{2}$ is on $\mathrm{C}_{m}$, and consists of two cells. One of the cells includes the left part of the cellular array that is denoted $\mathrm{C}_{1} \mathrm{C}_{m}$, and another includes the right part of the cellular array that is denoted $\mathrm{C}_{m+1} \mathrm{C}_{n}$. In this case, $\mathrm{G}_{2}$ of the right part is generated at $t=\frac{3 n-4}{2}$, and $\mathrm{G}_{2}$ of the left part is generated at $t=\frac{3 n-2}{2}$. Therefore the right part has been synchronized earlier than the left part. In this case, Waksman' solution is that the right part starts to synchronize with delayed 1 step. The technique is referred to as delaying. In the case that the parity of $\left|S_{i-1}\right|$ is an even number, $\mathrm{G}_{i}$ uses the delaying. When the parity of $\left|S_{i-1}\right|$ is an odd number, $\mathrm{G}_{i}$ doesn't use it.

## 4. Time-Optimum Firing Squad Synchronization Algorithm on $\mathbf{C A}_{1-\text { bit }}$

We design a firing squad synchronization algorithm on CA1-bit based on Waksman's algorithm.

### 4.1. Generation of infinite signals

The key idea is the following construction of an infinite set of 1-bit signals which propagate at $\frac{1}{2^{k+1}-1}-$ speed in one-way direction on a CA1-bit.
[Lemma 1] There exists a CA1-bit that can generate an infinite set of signals which are used efficiently in Waksman's algorithm [11]. Precisely, for any $n \geq 2$, the initial left-end General G generates $k$ signals $\mathrm{w}_{1}$, $\mathrm{w}_{2}, \ldots, \mathrm{w}_{k}$ propagating at speed $1 /\left(2^{k+1}-1\right)$ on $n$ cells, where $k=\left\lfloor\log _{2}(2 n-2)\right\rfloor-1$.

We need two bits that constitute the parity of $\left|S_{i-2}\right|$ and $\left|\mathrm{S}_{i-1}\right|$ so that we can generate $\mathrm{G}_{i}$ on right cell in real time. On the CA1-bit, both two bits can't be carried on a single 1-bit a-signal. Thus we have to develop a new technique for the real-time generation of generals.

### 4.2. Parity of $\left|\mathbf{S}_{i-2}\right|$

A a-signal, which propagates from $\mathrm{G}_{i}$ to $\mathrm{G}_{i+1}$, defines $\mathrm{a}_{\mathrm{G}_{i}}$-signal. The $\mathrm{a}_{\mathrm{G}_{i}}$-signal communicates a 1-bit information that is the parity of $\left|\mathrm{S}_{i-1}\right|$. This information is determined on $\mathrm{G}_{i}$, so that $\mathrm{a}_{\mathrm{G}_{i}}$-signal on CA 1 -bit can't communicate it like one on CA. And this information isn't communicated by $\mathrm{a}_{\mathrm{G}_{i}}$-signal, but it is


Figure 3: Generation of $\mathrm{G}_{3}$.


Figure 4: Relation of a-signal and $\mathrm{b}_{2}$-signal
determined by the cell on which $\mathrm{G}_{i+1}$ is generated. Therefore a-signal can't include this information. we need to find an other technique.
Here, we have found a rule, when $\mathrm{G}_{i}$ is generated by the crossing of $\mathrm{b}_{i}$-signal and a-signal. Figure 3 shows all cases in which $\mathrm{G}_{3}$ is generated on $\mathrm{C}_{7}$. In this figure, $n$ means $\left|\mathrm{S}_{0}\right| . \mathrm{G}_{3}$ is generated by the crossing of $\mathrm{b}_{3}$-signal and a-signal. We have investigated a parity of a cell on which $\mathrm{G}_{2}$, which has been the previous gerenal, has been located. As a result, when $G_{2}$ has been located at odd-numbered cell, which had been counted from $G_{0}, G_{3}$ has always consisted of only one cell. And when $\mathrm{G}_{2}$ has been located at even-numbered cell, which had been counted from $\mathrm{G}_{0}, \mathrm{G}_{3}$ has always consisted of two cells. And the parity of a cell, on which $G_{2}$ stays, is equal to the parity of $\left|S_{1}\right|$. If based on $G_{3},\left|S_{1}\right|$ becomes $\left|S_{i-2}\right|$.

Next, we must investigate how to get the parity of $\left|S_{i-2}\right|$ at $C_{7}$. The $b_{3}$-signal stays on $C_{7}$ between 7 steps. In this figure, a horizontal line is pulled to a half time of $b_{3}$-signal full staying time. If the crossing happens before the half time, $\mathrm{G}_{2}$ always positioned at odd-number cell. And if the crossing happens after the half time, $\mathrm{G}_{2}$ always positioned at even-number cell.

The $\mathrm{b}_{i}$-signal stays on $\mathrm{C}_{m_{i}}$ from $t=\left(2^{i}-1\right) m_{i}-2^{i}$ through $t=\left(2^{i}-1\right) m_{i}-2$. We define $\alpha_{i}$, however
$1 \leq \alpha_{i} \leq 2^{i}-1$ as a variable that means the offset time. We can express the staying time of $\mathrm{b}_{i}$-signal by using this equation.

$$
\begin{equation*}
t=\left(2^{i}-1\right) m_{i}-2^{i}-1+\alpha_{i} \tag{1}
\end{equation*}
$$

And, a-signal, which crosses $\mathrm{b}_{i}$-signal on $\mathrm{C}_{m_{i}}$, arrives at $\mathrm{C}_{m_{i}+1}$ at time $=t$, where $t$ is expressed by the following equation.

$$
\begin{equation*}
t=-m_{i}+2 n-2 \tag{2}
\end{equation*}
$$

From equations (1) and (2) we get:

$$
\begin{array}{r}
2^{i} m_{i}=2 n+2^{i}-1-\alpha_{i}  \tag{3}\\
1 \leq \alpha_{i} \leq 2^{i}-1
\end{array}
$$

$m_{i}$ is always an integer that leads to the equation(3), because $\alpha_{i}$ must be an odd number. When $\alpha_{i}$ is an even number, $\mathrm{b}_{i}$-signal doesn't cross the a-signal on $\mathrm{C}_{m_{i}}$. In addition, we are also examined that the crossings of $\mathrm{b}_{i-1}$-signal and a-signal on $\mathrm{C}_{m_{i-1}}$. We can also express $m_{i-1}$ by this equation such that:

$$
\begin{array}{r}
2^{i-1} m_{i-1}=2 n+2^{i-1}-1-\alpha_{i-1},  \tag{4}\\
1 \leq \alpha_{i-1} \leq 2^{i-1}-1
\end{array}
$$

Then we have equation(5).

$$
2^{i} \cdot m_{i}-2^{i-1} \cdot m_{i-1}=2^{i}-2^{i-1}-\alpha_{i}+\alpha_{i-1}(5)
$$

In this case, $\alpha_{i}$ and $\alpha_{i-1}$ are expressed as follows.
(i)If $\mathrm{C}_{m_{i-1}}$ is located on odd-number cell,

$$
\begin{equation*}
\alpha_{i}=\alpha_{i-1} \tag{6}
\end{equation*}
$$

(ii)If $\mathrm{C}_{m_{i-1}}$ is located on even-number cell,

$$
\begin{equation*}
\alpha_{i}=\alpha_{i-1}+2^{i-1} \tag{7}
\end{equation*}
$$

Therefore we get equations (8) and (9).
When $m_{i-1}$ is an odd number,

$$
\begin{equation*}
m_{i}=\frac{m_{i-1}+1}{2} \tag{8}
\end{equation*}
$$

And when $m_{i-1}$ is an even number,

$$
\begin{equation*}
m_{i}=\frac{m_{i-1}}{2} \tag{9}
\end{equation*}
$$

The $\mathrm{b}_{i}$-signal on CA 1 -bit needs a trigger signal that will arrive after a half time of $\mathrm{b}_{i}$-signal full staying time. Thus we have:
[Lemma 2] There exist a 1-bit signal that can determine the parity of $\left|\mathrm{S}_{i-2}\right|$ in time.

### 4.3. Parity of $\left|\mathbf{S}_{i-1}\right|$

The $\mathrm{a}_{\mathrm{G}_{i}}$-signal counts the parity of $\left|\mathrm{S}_{i}\right|$ from $\mathrm{G}_{i}$ to $\mathrm{G}_{i+1}$. The $\mathrm{a}_{\mathrm{G}_{i}}$-signal on CA1-bit can't count it. And this information is also determined by the cell on which $\mathrm{G}_{i+1}$ is generated. Therefore a-signal can't also include this information. we need to find an other technique. $\mathrm{G}_{i}$ stays on the center of $\mathrm{S}_{i-2}$. Here we define $\mathrm{C}_{1} \mathrm{C}_{\mathrm{G}_{i}}$, which is the left part of $\mathrm{S}_{i-2}$. Because $\mathrm{C}_{1} \mathrm{C}_{\mathrm{G}_{i}}$ is the same size as $S_{i-1}$, we can get the parity of $\left|S_{i-1}\right|$ at $\mathrm{C}_{\mathrm{G}_{i}}$, when $\mathrm{G}_{i}$ is generated. Figure 4 shows a propagation of the a-signal and the $\mathrm{b}_{i}$-signal in Waksman's algorithm. In this figure, we focus an offset time after a-signal passes until $\mathrm{b}_{2}$-signal arrives. On any $\mathrm{C}_{m}$, where $m$ is a positive integer and it is bigger than 1 , the offset time can lead to this equation.

$$
\begin{equation*}
\Delta t=2 p-3 \tag{10}
\end{equation*}
$$

In this case, we define $\Delta t$ that is the remainder which divided the offset time by 4 , If $p=2 x$,

$$
\begin{equation*}
\Delta t \bmod 4=4 x-3=4(x-1)+1 \tag{11}
\end{equation*}
$$

If $\quad p=2 x+1$,

$$
\begin{equation*}
\Delta t \bmod 4=4 x-1=4(x-1)+3 \tag{12}
\end{equation*}
$$

[Lemma 3] There exist 1-bit signals that can determine the parity of $\left|S_{i-1}\right|$.

Based on lemmas above, our main theorem is stated as follows:
[Theorem 4] There exists a CA1-bit which can synchronize any $n$ cells in $2 n-2$ steps. A CA1-bit implemented has 78 internal states and 208 transition rules.

In Fig. 5 we show snapshots of the synchronization processes. Small right and left black triangles, shown - and $\boldsymbol{\iota}$ in the figure, indicate a 1-bit transfer in the right or left direction between neighboring cells. A symbol in a cell shows its internal state.

## 5. Conclusion

We have designed and implemented an optimumtime firing squad synchronization algorithm on $\mathrm{CA}_{1}$ bit. Each cell has 78 internal states and 208 transition rules, and we checked its validity from $n=2$ through $n=10000$ by computer simulation.

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Figure 5: Snapshots of our time-optimum synchronization on $\mathrm{CA}_{1 \text {-bit }}$ consisting of 25 cells.
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